













APPENDIX A  
PROOF OF PROPOSITION

It is described,  $X_v = |\sum_{l=1}^N \hat{h}_{vl}^H \Phi \hat{g}_{vl}|^2$ . In the event that it is thought that the signal reflected by the RIS components contains random phase fluctuations. It is acceptable to suppose that  $\hat{h}_l \sim \mathcal{CN}(0, \beta_{SR} I_N)$ ,  $\hat{g}_{vl} \sim \mathcal{CN}(0, \beta_{RU_v} I_N)$  and  $\Phi = \text{diag}(e^{j\theta_1}, e^{j\theta_2}, \dots, e^{j\theta_N})$ , where  $I_N$  and  $\theta_n \in [-\pi, \pi]$  are character matrices of order  $N$  with a range for  $\theta_n$ , respectively, due to the identical independent distribution of each channel (i.i.d). In addition, we define the anticipated value. The independence of the channels that make up  $\mathbb{E}\{X_v\}$  are averaged. Eq could be used to express it (A.1).

$$\mathbb{E}\{X_v\} = \mathbb{E}\left\{|\hat{h}_l^H \Phi \hat{g}_{vl}|^2\right\} \quad (\text{A.1})$$

$X_v$  can be written as an equation and represents the average link  $\hat{X}_v$  and fade due to path-loss distance-induced attenuation  $\sigma_{e_v}^2$  in the propagation location (A.2).

$$\mathbb{E}\{X_v\} = \mathbb{E}\{\hat{X}_v + \sigma_{e_v}^2\} = \mathbb{E}\{\hat{X}_v\} + \sigma_{e_v}^2 \quad (\text{A.2})$$

Path-loss distance-induced attenuation  $\sigma_{e_v}^2$  could be defined as  $\frac{\eta_{SR}}{N} (d_{SR}^{-\chi}) \cdot \frac{\eta_{RU_v}}{N} (d_{RU_v}^{-\chi})$ , then Eq. (A.2) might be revised as Eq. (A.3).

$$\mathbb{E}\{X_v\} = \mathbb{E}\left\{|\hat{h}_l^H \Phi \hat{g}_{vl}|^2\right\} + \frac{\eta_{SR}}{N} (d_{SR}^{-\chi}) \cdot \frac{\eta_{RU_v}}{N} (d_{RU_v}^{-\chi}) \quad (\text{A.3})$$

We define that  $\mathbb{E}\left\{|\hat{h}_l^H \Phi \hat{g}_{vl}|^2\right\} = N\beta_{SR}\beta_{RU_v}$ , so that Eq. (A.3) could be written as Eq. (A.4)

$$\mathbb{E}\{X_v\} = N\beta_{SR}\beta_{RU_v} + \frac{\eta_{SR}}{N} (d_{SR}^{-\chi}) \cdot \frac{\eta_{RU_v}}{N} (d_{RU_v}^{-\chi}) \quad (\text{A.4})$$

Furthermore, it is determined  $\mathbb{E}\{|\hat{X}_v|^2\}$  as Eq. (A.5) below.

$$\mathbb{E}\{|\hat{X}_v|^2\} = \mathbb{E}\left\{|\hat{h}_l^H \Phi \hat{g}_{vl}|^4\right\} \quad (\text{A.5})$$

Using the circular symmetric properties, Eq. (A.5) could be written as Eq. (A.6).

$$\mathbb{E}\{|\hat{X}_v|^2\} = \mathbb{E}\left\{|\Phi \hat{g}_{vl}|^4 \left|\frac{\hat{h}_l^H \Phi \hat{g}_{vl} \hat{g}_{vl}^H \Phi^H \hat{h}_l}{\|\Phi \hat{g}_{vl}\| \|\Phi \hat{g}_{vl}\|}\right|^2\right\} \quad (\text{A.6})$$

If it is defined  $z = \frac{\hat{h}_l^H \Phi \hat{g}_{vl}}{\|\Phi \hat{g}_{vl}\|}$ ;  $z \sim \mathcal{CN}(0, \beta_{SR})$ , then

$$\mathbb{E}\{|\hat{X}_v|^2\} = \mathbb{E}\left\{|\Phi \hat{g}_{vl}|^4 |z|^4\right\} = \mathbb{E}\left\{|\Phi \hat{g}_{vl}|^4\right\} \mathbb{E}\{|z|^4\} \quad (\text{A.7})$$

We could get Eq. (A.24) by entering the values  $\mathbb{E}\left\{|\Phi \hat{g}_{vl}|^4\right\} \mathbb{E}\{|z|^4\}$  into Eq (A.6).

It can also be described as  $X_v = \hat{X}_v + e_v = \hat{X}_v + \sigma_{e_v}^2$ , allowing the expected-value of  $X_v^2$  to express in the form of Eq. (A.8).

$$\mathbb{E}\{X_v^2\} = N(N+1)\beta_{RU_v}^2\beta_{SR}^2 = (2N^2 + 2N)\beta_{RU_v}^2\beta_{SR}^2 \quad (\text{A.8})$$

Based on Eq. (A.8), we specify  $\mathbb{E}\{X_v^2\} = \mathbb{E}\{|\hat{X}_v|^2\} + \sigma_{e_v}^2$  and could be rewritten as Eq. (A.9) ad Eq. (A.10).

$$\mathbb{E}\{X_v^2\} = \mathbb{E}\left\{|\hat{h}_l^H \Phi \hat{g}_{vl}|^4 + \left|\frac{\eta_{SR}}{N} (d_{SR}^{-\chi}) \cdot \frac{\eta_{RU_v}}{N} (d_{RU_v}^{-\chi})\right|^2\right\} \quad (\text{A.9})$$

$$\mathbb{E}\{X_v^2\} = (2N^2 + 2N)\beta_{RU_v}^2\beta_{SR}^2 + \left|\frac{\eta_{SR}}{N} (d_{SR}^{-\chi}) \cdot \frac{\eta_{RU_v}}{N} (d_{RU_v}^{-\chi})\right|^2 \quad (\text{A.10})$$

By defining,  $\text{Var}\{X_v\} = E\{X_v^2\} - |E\{X_v\}|^2$ , then it may be said to be as Eq.(A.11).

$$\text{Var}\{X_v\} = (N^2 + 2N)\beta_{RU_v}^2\beta_{SR}^2 - 2N\beta_{SR}\beta_{RU_v} \frac{\eta_{SR}}{N} (d_{SR}^{-\chi}) \cdot \frac{\eta_{RU_v}}{N} (d_{RU_v}^{-\chi}) \quad (\text{A.11})$$

Next,  $\tau_1$  in Eq. (11) could be derived by defining  $\rho_{Th_1}$  as Eq. (A.12).

$$\rho_{Th_1} = \frac{(2^{2\zeta_1-1})}{\rho_s} \quad (\text{A.12})$$

From Eq. (7), we define the notation  $A_1 = |\sum_{l=1}^L \hat{h}_l| |\hat{g}_{1l}|^2$  and  $\lambda_1 = \frac{\eta_{SR}}{N} (d_{SR}^{-\chi}) \cdot \frac{\eta_{RU_1}}{N} (d_{RU_1}^{-\chi}) \rho_s + 1$  so that it can be rewritten Eq. (7) as Eq. (A.18).

$$\frac{A_1 a_1 \rho_s}{A_1 a_1 \rho_s + \lambda_1} \geq \rho_{Th_1} \quad (\text{A.13})$$

$$\Rightarrow A_1 a_1 \rho_s \geq (A_1 a_1 \rho_s + \lambda_1) \rho_{Th_1} \quad (\text{A.14})$$

$$\Rightarrow A_1 a_1 \rho_s - A_1 a_1 \rho_s \rho_{Th_1} \geq \lambda_1 \rho_{Th_1} \quad (\text{A.15})$$

$$\Rightarrow A_1 \rho_s (a_1 - a_1 \rho_{Th_1}) \geq \lambda_1 \rho_{Th_1} \quad (\text{A.16})$$

$$\Rightarrow A_1 \geq \frac{\lambda_1 \rho_{Th_1}}{(a_1 - a_1 \rho_{Th_1}) \rho_s} \quad (\text{A.17})$$

$$\Rightarrow \tau_1 = \frac{\lambda_1 \rho_{Th_1}}{(a_1 - a_1 \rho_{Th_1}) \rho_s} \quad (\text{A.18})$$

Similarly, it is defined  $\rho_{Th_2}$  in Eq. (A.19).

$$\rho_{Th_2} = \frac{(2^{2\zeta_2-1})}{\rho_s} \quad (\text{A.19})$$

then,  $\tau_2$  could be derived from Eq. (A.19) as follows.

$$\frac{A_2 a_2 \rho_s}{A_2 a_2 \rho_s + \lambda_1} \geq \rho_{Th_2} \quad (\text{A.20})$$

$$\Rightarrow A_2 a_2 \rho_s \geq (A_2 a_2 \rho_s + \lambda_1) \rho_{Th_2} \quad (\text{A.21})$$

$$\Rightarrow A_2 a_2 \rho_s - A_2 a_2 \rho_s \rho_{Th_2} \geq \lambda_1 \rho_{Th_2} \quad (\text{A.22})$$

$$\Rightarrow A_2 \rho_s (a_2 - a_2 \rho_{Th_2}) \geq \lambda_1 \rho_{Th_2} \quad (\text{A.23})$$

$$\Rightarrow A_2 \geq \frac{\lambda_1 \rho_{Th_2}}{(a_2 - a_2 \rho_{Th_2}) \rho_s} \quad (\text{A.24})$$

$$\Rightarrow \tau_2 = \frac{\lambda_1 \rho_{Th_2}}{(a_2 - a_2 \rho_{Th_2}) \rho_s} \quad (\text{A.25})$$

The outage probability at user  $U_I$  could be mentioned in Eq. (A.26), which is determined by making changes to the equations of PDF and CDF given above as a result of the deployment of RIS in NOMA networks.

$$P_{U_1} = 1 - \Pr(|\hat{X}_1|^2 > \tau_2, |\hat{X}_1|^2 > \tau_1) \quad (\text{A.26})$$

so that, by rearranging Eq. (A.26) obtained as Eq. (A.27).

$$P_{U_1} = 1 - \Pr(|\hat{X}_1|^2 > \max(\tau_2, \tau_1)) \quad (\text{A.27})$$

$$P_{U_1} = 1 - (|\hat{X}_1|^2 > \tau) \quad (\text{A.28})$$

where, according to which value is the highest,  $\tau$  might be as high as  $\tau_2$  or  $\tau_1$ . The reason for this is that there is a higher probability that  $U_1$  will not successfully decode the signal  $x_2$  from another user,  $U_2$ , or its signal,  $x_2$ , than that any of those two events will occur. Consequently, Eq. (A.27) is written as Eq. (A.28).

APPENDIX B  
PROOF OF PROPOSITION

According to the equation of  $\gamma_{U_2,BS}$ , in the Eq. (B.1).

$$\gamma_{U_2,BS} = \frac{|h_{D_2}|^2 a_2 \rho_s}{a_1 \rho_s + \eta_{SU_2} d_{SU_2}^{-\chi} \rho_s + 1} \quad (\text{B.1})$$

We derive  $\tau_4$  base on the system model, which could be conducted as follows.

If it is defined  $B = |\hat{h}_{D_2}|^2$  and  $\lambda_3 = \eta_{SU_2} (d_{SU_2}^{-\chi}) \rho_s + 1$ , then

$$\Rightarrow \frac{|h_{D_2}|^2 a_2 \rho_s}{|\hat{h}_{D_2}|^2 a_1 \rho_s + \eta_{SU_2} (d_{SU_2}^{-\chi}) \rho_s + 1} \geq \rho_{Th_2} \quad (\text{B.2})$$

$$\Rightarrow \frac{B a_2 \rho_s}{B a_1 \rho_s + \lambda_3} \geq \rho_{Th_2} \quad (\text{B.3})$$

$$\Rightarrow B a_2 \rho_s \geq (B a_1 \rho_s + \lambda_3) \rho_{Th_2} \quad (\text{B.4})$$

$$\Rightarrow B a_2 \rho_s - B a_1 \rho_s \rho_{Th_2} \geq \lambda_3 \rho_{Th_2} \quad (\text{B.5})$$

$$\Rightarrow B \rho_s (a_2 - a_1 \rho_{Th_2}) \geq \lambda_3 \rho_{Th_2} \quad (\text{B.6})$$

$$\Rightarrow B \geq \frac{\lambda_3 \rho_{Th_2}}{(a_2 - a_1 \rho_{Th_2}) \rho_s} \quad (\text{B.7})$$

$$\Rightarrow \tau_4 = \frac{\lambda_3 \rho_{Th_2}}{(a_2 - a_1 \rho_{Th_2}) \rho_s} \quad (\text{B.8})$$

The outage probability at user  $U_2$  could be shown as follows.

$$P_{U_2} = Pr(\gamma_{U_2,BS} < \gamma_{Th_2}) + Pr(\gamma_{U_2,BS} > \gamma_{Th_2}, \gamma_{U_2,U_1} < \gamma_{Th_2}) \quad (\text{B.9})$$

$$P_{U_2} = Pr(\gamma_{U_2,BS} < \gamma_{Th_2}) + Pr(\gamma_{U_2,BS} > \gamma_{Th_2}) Pr(\gamma_{U_2,U_1} < \gamma_{Th_2}) \quad (\text{B.10})$$

$$P_{U_2} = 1 - Pr(\gamma_{U_2,BS} > \gamma_{Th_2}) Pr(\gamma_{U_2,U_1} > \gamma_{Th_2}) \quad (\text{B.11})$$

$$P_{U_2} = 1 - e^{-\delta_1 \tau_2} \sum_{j=0}^{m_1-1} \frac{(\delta_1 \tau_2)^j}{j!} e^{-\delta_3 \tau_4} \sum_{k=0}^{m_3-1} \frac{(\delta_3 \tau_4)^k}{k!} \quad (\text{B.12})$$

NOTATION  
TABLE OF NOTATION [35].

Notation	Definition
$s(t)$	A superimposed signal is sent to both the near ( $U_1$ ) and far ( $U_2$ ) users.
$P_s$	Transmitted signal power
$P_1$	Transmitted signal power by $U_1$
$a_1$ and $a_2$	Power level $x_1$ and $x_2$
$N$	The quantity of RIS components

Notation	Definition
$\alpha$	The RIS Reflected Signal's Amplitude Coefficient with $\alpha \in (0,1]$
$\theta_l$	Phase modification is possible with the $l$ -th reflected element of RIS.
$\Phi$	The <i>phase-shift matrix</i> , $\text{diag}(\exp(j\theta_1), \exp(j\theta_2), \dots, \exp(j\theta_L))$
$(\cdot)^H$	Hermitian transpose
$\beta_k$	The large-scale fading coefficients of the channel $k$
$\Omega_k$	The link power of channel $k$
$\hat{\Omega}_k$	Average-connection power of channel $k$
$\hat{h}_k$	Average-fading coefficient of channel $k$
$h_k$	Fading-coefficient of channel $k$
$\hat{\chi}_v$	Average fading-coefficient gain by RIS of $v$ user
$X_v$	Gain in fading coefficient by $v$ user's RIS
$e_k$	Channel estimation error
$\sigma_{e_k}^2$	Channel estimation error variant
$\eta_k$	Channel $k$ with relative channel estimation error
$m_v$	The gamma distribution channel's shape factor
$\chi$	Path-loss exponent
$d_k$	Distance of two-points crossed by channel $k$
$P_{(\cdot)}$	Outage probability at $(\cdot)$
$n_{U_1}$ and $n_{U_2}$	AWGN at $U_1$ and $U_2$ .
$d_{SU_1}, d_{SR_1}, d_{RU_1}$	Distance of BS – $U_1$ , BS – RIS, and RIS – $U_1$
$d_{SU_2}, d_{SR_2}, d_{RU_2}$	Distance of BS – $U_2$ , BS – RIS, and RIS – $U_2$
$h_{D_1}, h_{D_2}, h_l, g_{1l}, g_{2l}$	Coefficients of fading channel
$\rho_s$	Transmit signal-to-noise ratio (SNR)
$\rho_{U_2 \rightarrow U_1}$	At $U_1$ , examine the incoming signal for interference and noise ratio in order to decode it (SINR).
$\rho_{U_1}, \rho_{U_2}$	The received SINR of $U_1$ and $U_2$ to decode the signal itself.
$\rho_{2,U_2}$	The received SINR of $U_2$ to decode the signal $x_2$ to relay link
$\rho_{U_2}^{SC}$	The received SINR after selection combining (SC) at $U_2$
$\rho_{Th_1}$ and $\rho_{Th_2}$	SINR target of user $U_1$ and $U_2$
$R_1$ and $R_2$	Target rate of user $U_1$ and $U_2$
$P_{U_1}$ and $P_{U_2}$	Outage probability of $U_1$ and $U_2$
$\tau_1, \tau_2$ and $\tau_3$	The first-comparison parameter, the second-comparison parameter, and the third-comparison parameter are all variables.
$\lambda_1, \lambda_2$	Interference and noise due to the using of RIS-aided of $U_1$ and $U_2$
$\delta_1, \delta_2$	Scale factor of the gamma-distribution of the channel of $U_1$ and $U_2$